

IETE Technical Review



ISSN: 0256-4602 (Print) 0974-5971 (Online) Journal homepage: http://www.tandfonline.com/loi/titr20

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To cite this article: Cheng Jin & Sung Ho Cho (2013) An Enhanced Collision Detection Scheme for ISO 18000–6C RFID Systems, IETE Technical Review, 30:2, 108-119

To link to this article: https://doi.org/10.4103/0256-4602.110549



An Enhanced Collision Detection Scheme for ISO 18000-6C RFID Systems

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Abstract

In radio-frequency identification systems, tag collision resolution is a significant issue for fast tag identification. Dynamic framed slotted ALOHA (DFSA) is one of the most widely used algorithms to resolve tag collision. Collision detection (CD) plays an important role in determining the efficiency of DFSA-based algorithms because most DFSA-based algorithms determine the next frame size according to the number of collided slots in the current frame. Existing CD methods do not respond quickly enough to detect a collision and have difficulty in distinguishing a collision from noise, resulting in a degradation of system efficiency. This paper presents a CD scheme based on the ISO/IEC 18000-6C protocol to improve CD efficiency. This scheme enables fast and accurate CD by detecting the number of pulses transmitted by tags. The effectiveness and practical feasibility of the scheme is verified by simulation and implementation. Performance evaluation results show that the proposed scheme achieves faster identification speed than the conventional methods, especially under noise conditions.

Keywords

Access protocols, ALOHA, Collision detection, Gen2, Radio-frequency identification, Tag collision.

1. Introduction

Radio-frequency identification (RFID) is an automatic wireless identification technology which has become an attractive solution for object tracking [1,2], supply chain management and industry automation. An RFID system generally consists of an RFID reader and one or more RFID tags, which are typically attached to objects that need to be identified by the reader. To identify tags, the reader sends query commands to the tags and the tags send response back to the reader.

According to the collision problems in RFID systems, they can be classified into the following two types: Reader collision and tag collision. Reader collision occurs in a multiple reader environment [3] while tag collision occurs in a single reader environment when the reader attempts to identify a large number of tags. In this paper, we focus on the tag collision problem. One primary goal of RFID systems is to identify multiple tags as fast as possible. However, if more than one tag replies to the reader simultaneously, a tag collision occurs and none of the tags can be identified by the reader. Therefore, collision resolution is a critical issue in determining tag identification speed. To resolve the collision problem, many anti-collision algorithms have been presented. There are two main kinds of anti-collision algorithms used in RFID systems: Tree-based [4-10] and ALOHA-based [11-17]. The tree-based algorithm splits the collided tags into subsets when a collision occurs, then proceeds each subset separately. The process is repeated until all tags are identified. The tree-based algorithm may

require much time when the number of tags is large. In ALOHA-based algorithm, the reader divides time into intervals of time slots and each tag randomly selects one time slot and sends its ID in that time slot. Generally, slots are issued in frames (from which the name framed slotted ALOHA), the size of which is determined by the reader at the beginning of each frame. In a frame, if two or more tags select a same slot to transmit data, a collision occurs and the collided tags retransmit in the next frame. The performance of framed slotted ALOHA (FSA) depends on the number of tags and the frame size. The maximum efficiency of FSA can be achieved when the frame size is equal to the number of tags [18]. Dynamic framed slotted ALOHA (DFSA) is an advanced version of FSA which dynamically changes the frame size according to the number of unidentified tags to optimize the efficiency of FSA. In DFSA, in order to determine the optimal frame size for the next frame, the number of unidentified tags needs to be estimated at the current frame. Therefore, estimating the number of unidentified tags plays a crucial role in determining the efficiency of DFSA.

Recently, many works on tag estimation method have been proposed to improve the performance of DFSA-based RFID systems [19-22]. In these methods, tag estimation is based on observing the number of empty, successful and collided slots in a frame. A common feature of these methods is that they operate under assumption that the information of slot occupancy is error-free. However, in practical situations, this assumption is not always true due to reasons such as noise. The

decision on slot occupancy—whether the slot contains 0, 1, or more replies, directly determines the number of the empty slots, the successful slots and the collided slots. Therefore, collision detection (CD), detecting a collision and distinguishing it from empty and successful slots, is an important factor in determining the overall system performance.

In passive RFID systems, the conventional methods for detecting a tag collision are mainly achieved by detecting a signal corruption, or analyzing the signal waveform [23-26]. Khasgiwale et al. [23] presented a CD technique that detects the number of tags involved in a collision by analyzing the radar cross-section (RCS) state in the tag signal waveform. This technique enables detection of up to 4 collided tags, however, requires support of professional tools, like the NI-PXI hardware for signal capture and LabVIEW software for signal processing. Liu et al. [24] proposed a CD method by detecting a corruption in the received signal based on the coding characteristics of the signal. Yang et al. [25] proposed a quick CD approach based on checking a newly designed preamble in the tag signal, which is supposed to be implemented on a tag. Kang et al. [26] designed a collision decoding scheme based on a CD performed by detecting the signal length. The drawbacks of these methods are two-fold: (1) A relatively long time required wasting slot duration time; (2) When signal corruption is detected, it is difficult to determine whether it is caused by tag collision or just by noise such that the number of unnecessary slots may be increased. These problems can cause the overall identification speed degraded. Thus, to overcome these problems, we propose a CD scheme based on the ISO/IEC 18000-6C RFID protocol, the anti-collision algorithm of which is a kind of DFSA. In the proposed scheme, CD is achieved by detecting the number of pulses transmitted by tags to detect a collision faster and more accurately.

The remainder of this paper is organized as follows. In Section 2, we describe the operation of the Gen2

RFID system. In Section 3, we describe the proposed CD scheme and analyze its performance. In Section 4, performance evaluation results of the proposed scheme obtained by simulation are presented. In Section 5, we discuss the implementation issues of the proposed scheme. In Section 6, we discuss the practical issue about how to separately identify the proposed tags from conventional tags when they are mixed together. In Section 7, the conclusions are drawn.

2. System Description

Our proposed scheme is for the RFID systems based on the ISO/IEC 18000-6C protocol, which is also known as the EPCglobal Class-1 Generation-2 protocol (hereafter referred to as the Gen2 protocol). This protocol provides superior new features including flexibility, robustness and fast tag identification. A full specification of the Gen2 protocol can be found in [27]. In the Gen2 protocol, a *Q* algorithm, which is a variant of the DFSA algorithm, is proposed as the anti-collision algorithm [27].

We use Figure 1 to describe the tag identification procedure using the *Q* algorithm in the Gen2 protocol. Firstly, the reader issues a Query command which contains a Q parameter to set the frame size as 2^Q. Then, each tag receiving the Query command randomly selects a slot number (SN) in the range of $[0, 2^{Q}-1]$. Only the tag whose SN is zero responds to the reader by sending an RN16, a 16-bit random number, within time T_1 . If the reader successfully receives the RN16, it sends back an ACK command which contains the same RN16 within time T_2 . If the tag's RN16 is the same with that in the ACK command, it sends its ID, labeled EPC in Figure 1, to the reader. On the other hand, the tag whose SN is other than 0 waits a QueryRep command from the reader to cause its SN to decrement by 1. QueryRep means "repeat query"; this command is used to cause all tags to decrement their SN by 1.

This process continues until the SN becomes 0, after which the procedure for the tag whose SN is zero is

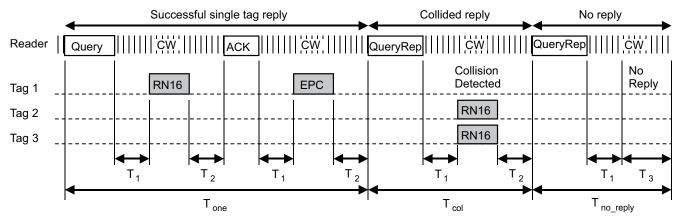


Figure 1: Tag identification procedure in Gen2 protocol.

followed. Since each tag performs the same procedure independently, there are three possible situations after the reader issues a query command (Query, QueryRep, or QueryAdjust) to tags:

- Successful Single Tag Reply: Only one tag responds to the reader and the reader successfully receives the RN16. Then, the reader acknowledges the tag by sending an ACK. If the tag receives the ACK with a correct RN16, it sends its EPC
- 2. Collided Reply: Multiple tags respond simultaneously and a collision occurs. The reader observes a waveform comprising multiple RN16s. The reader issues a QueryRep or QueryAdjust command to resolve the collision and continue to identify tags. The QueryAdjust command is used to adjust the value of *Q* and cause all tags to re-select their SN based on the new *Q*. The value of *Q* could be incremented by 1, decremented by 1, or remained unchanged according to the *Q* algorithm
- 3. No Reply: No tag responds to the reader. The reader adjusts the value of *Q* according to the *Q* algorithm and continues to identify tags by sending a QueryRep or QueryAdjust command after a waiting time, *T*₃. The tags that has SN other than 0 receive the QueryRep command and decrement their SN by 1. This process continues until the SN reaches 0, after which the procedure in (1) will be followed.

3. Proposed Collision Detection Scheme

We propose a scheme to improve the CD efficiency. The proposed scheme is based on the idea of carrier sense multiple access with time-split collision detection (CSMA/TCD) first introduced to enable CD in radio channels [28]. In CSMA/TCD, each terminal transmits a preamble before data transmission and detects collided transmission from other terminals by carrier sensing. However, in passive RFID systems, since tags are very simple devices without the ability to perform carrier sensing, tags cannot detect or avoid collisions with each other. It is up to the reader to detect and resolve tag collisions. Basing on these considerations, we proposed our CD scheme.

3.1 Description of the Proposed Scheme

The operation of the proposed CD scheme is illustrated in Figure 2. In the proposed scheme, each tag transmits a pulse with a length of T_{vri} in a random selected time slot whose period is T_{n} , among total L time slots before transmitting an RN16. The slot selection is scheduled by controlled randomization process, slotted-ALOHA method. In addition, considering the propagation delay and to guarantee detection of separate individual pulses, each pulse transmission is preceded by an idle period with an equal length to the pulse. In the return link, the reader monitors the pulses transmitted by tags: Whether there is zero, one, or more pulses. A collision will be indicated if more than one pulse is detected. Using this method, CD can be performed faster and more accurately in the following two ways: (1) The time required to determine if there is a collision is reduced from a full RN16 transmission time to the pulse detection time, Δt . Since the slot time *T*_n for one pulse transmission is less than 1-bit

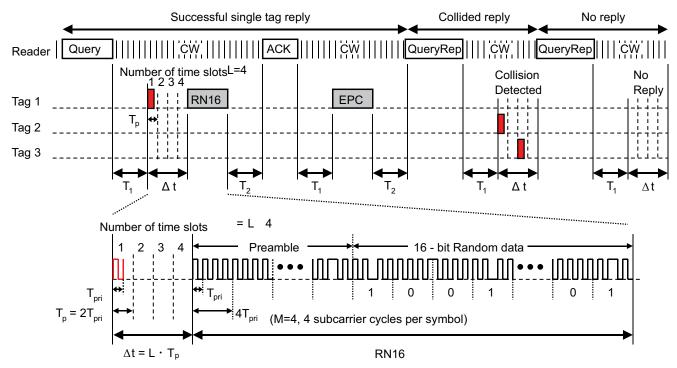


Figure 2: Tag identification procedure with proposed CD scheme.

(2)

period in the RN16, the pulse detection time Δt is very short compared with the RN16 frame ($\Delta t = 5\%~T_{RN16}$ for L=4 in Figure 2). (2) By detecting the number of pulses to determine a collision, a collision can be unambiguously distinguished from a single tag reply corrupted by noise.

Aiming at maximizing detectability while minimizing implementation complexity, the pulse waveform is proposed to be the same as one subcarrier cycle in the RN16, the period of which is $T_{pri'}$ as shown in Figure 2. For pulse detection, a square-law detector is proposed at the reader receiver for its effectiveness and design simplicity.

3.2 Analysis of the Proposed Scheme

In this section, we analyze and compute the optimal performance that can be achieved using the proposed scheme. The performance metric we consider is the total tag identification time. According to the Gen2 system, the total time to identify n_1 tags in error-free link can be calculated as

$$t = n_0 T_{no \ revly} + n_1 T_{one} + n_2 T_{col}$$
 (1)

where, n_0 , n_1 and n_2 are the occurrence times of No Reply, Single Tag Reply and Collided Reply shown in Figure 1, respectively. Accordingly, T_{no_reply} , T_{one} and T_{col} are the time durations for each case above and can be calculated as

$$\begin{split} T_{no_reply} &= T_{QueryRep} \, + T_1 + T_3 \,, \\ T_{one} &= T_{QueryRep} \, + T_1 + T_{RN\,16} + T_2 + T_{ACK} \, + T_1 + T_{EPC} \, + T_2 \\ T_{col} &= T_{QueryRep} \, + T_1 + T_{RN\,16} \, + T_2 \end{split}$$

Note that T_{col} in (2) is the time duration for Collided Reply using the conventional CD method in [24], as shown in Figure 1. For the proposed CD scheme, as can be seen from Figure 2, the total time to identify the same number of tags can be calculated as

$$\begin{split} t_{CD} &= n_0 (T_{no_reply} - T_3 + \Delta t) + n_1 (T_{one} + \Delta t) \\ &+ n_2 \eta_{CD} (T_{no_reply} - T_3 + \Delta t) \\ &+ n_2 (1 - \eta_{CD}) (T_{col} + \Delta t), \end{split} \tag{3}$$

where, $\Delta t = L \cdot T_p$ [Figure 2] is the processing time of CD and η_{CD} is the average success probability of CD, which we will analyze in detail in the following.

We propose that pulse transmission be scheduled by a controlled randomization process—the slotted ALOHA method. Since each pulse is transmitted in random selection, the equal length time slot ahead of RN16, a situation exists where all pulses are distributed into a complete same time slot. In this situation, the multiple pulses will be detected as only one pulse, such that the collision will go undetected. As a result, a wrong decision "Single Tag

Reply" instead of "Collided Reply" will be made. This situation caused by the complete same slot occupancy is considered as a CD failed case in the proposed scheme. We define p_{CD_fail} as the failed probability of a CD for the situation above. Supposing n pulses, transmitted by n tags, are randomly distributed into L time slots, the probability that k pulses are distributed into one time slot can be calculated as [29]

$$p_{k} = \binom{n}{k} \left(\frac{1}{L}\right)^{k-1} \left(1 - \frac{1}{L}\right)^{n-k}$$
 (4)

Let k = n, then we get the probability that all pulses are distributed into the same slot, which is also the p_{CD_fail} we defined above

$$p_{CD \ fail} = p_{k=n} = (1/L)^{n-1} \tag{5}$$

Therefore, the success probability of a CD is

$$p_{CD_succ} = 1 - p_{CD_fail} = 1 - (1/L)^{n-1}$$
(6)

We use η_{CD} in (3) to describe how many times the CD has succeeded in a complete identification process. η_{CD} can be calculated as the ratio of successful CD times n_{CD_succ} over all collision occurrence times N

$$\eta_{CD} = \frac{n_{CD_succ}}{N} \tag{7}$$

Among the total N collision occurrences, in each collision, there could be 2, 3, or more tags involved in the collision. Thus, we can define c_k as the number of collisions which involve k tags in an inventory frame, the expected value of c_k can be calculated as

$$c_k = L \cdot p_{k'} \tag{8}$$

where, L is the number of time slots in the frame and p_k is the probability that k tags are involved in the same slot, which has been defined in (4). Then, we have

$$N = \sum_{k=2}^{M} c_k, \tag{9}$$

where, M is the maximum number of collided tags which are involved in a collision. Therefore, the expected value of $\eta_{CD~succ}$ can be calculated as

$$n_{CD_succ} = \sum_{k=2}^{M} c_k \cdot p_{CD_succ}(k) = \sum_{k=2}^{M} c_k \cdot (1 - (1/L)^{k-1})$$
 (10)

Substituting (9) and (10) into (7), the CD success probability η_{CD} becomes

$$\eta_{CD} = \frac{\sum_{k=2}^{M} c_k \cdot (1 - (1/L)^{k-1})}{\sum_{k=2}^{M} c_k}$$
 (11)

Let $Norm_k = c_k / \Sigma_{k=2}^M c_k$, then $Norm_k$ denotes the normalized occurrence rate of the k-tag involved collision. Finally, (11) becomes

$$\eta_{CD} = \sum_{k=2}^{M} Norm_{k} \cdot (1 - (1/L)^{k-1})$$
(12)

By dynamically adjusting the frame size L according to the Q algorithm to identify a population of tags, $Norm_k$ can be computed. A calculation result of $Norm_k$ is shown in Table 1, obtained by identifying 200 tags (X = 200) with an initial frame size, L = 16. When the number of tags X becomes large, $Norm_k$ shows a little difference accordingly, since the maximum value of k increases.

Based on Table 1, a plot of $Norm_k$ for different k is shown in Figure 3. It can be seen that the distribution of $Norm_k$ is close to a Poisson distribution [18] but not exactly the same. The reason could be that the frame length cannot always be exactly equal to the number of unidentified tags in the Q algorithm.

In addition, using the same derivation for c_k in (8), when identifying n_1 tags, we can obtain the expected values of No Reply n_0 and Collided Reply n_2 , which have been defined in (1). Due to limitation of space, the derivation is not shown. They can be approximated by

$$n_0 \approx 1.60 n_1, n_2 \approx 0.93 n_1$$
 (13)

From (13), we could get a system efficiency of $n_1/(n_0 + n_1 + n_2) = 28\%$, which matches the analytical result in [30].

Figure 4 shows a plot of the analytical calculation of η_{CD} for different time slots, L. The calculation is based on (12) and Table 1. As shown in Figure 4, η_{CD} increases as the number of time slots L increases. However, the processing time of CD also increases since the processing time is $\Delta t = L \cdot T_p$, in which T_p is the period of one time slot for pulse transmission [Figure 2]. Therefore, it is necessary to find out the optimal number of time slots— L_{op} , which minimizes the total identification time t_{CD} .

For the proposed CD scheme, the total identification time reduction— Δt_{CD} , compared with the conventional CD method in (1), can be calculated as

$$\begin{split} \Delta t_{CD} &= t_{CD} - t \\ &= (n_0 + n_1 + n_2) \Delta t - n_2 \eta_{CD} \left(T_{col} - T_{no_reply} + T_3 \right) - n_0 T_3 \\ &\approx n_1 [3.53 L \cdot T_p - 0.93 \eta_{CD} \left(T_{col} - T_{no_reply} + T_3 \right) - 1.60 T_3] \end{split}$$
 (14)

Table 1: Calculated norm, for k (identify 200 tags)

k	Norm _k	k	Norm _k
2	0.4714	6	0.0514
3	0.2192	7	0.0352
4	0.1222	8	0.0235
5	0.0767		

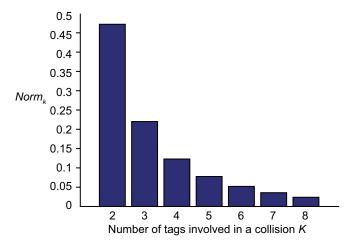


Figure 3: Normalized collision occurrence rate $Norm_k$ for k-tag involved collision (using results in Table 1).

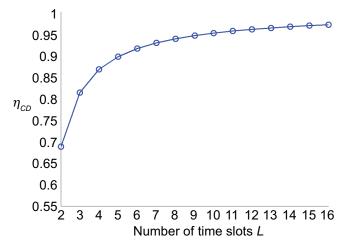


Figure 4: CD Success probability $\eta_{\rm CvD}$ according to number of time slots L.

Our goal is to find $L = L_{op}$ such that t_{CD} is minimized. Thus, the optimization problem can be stated as

$$L_{op} = \arg \max_{I} |\Delta t_{CD}| \tag{15}$$

To solve this, we examine the parameters in (14) in detail. Table 2 gives the typical system parameters of the Gen2 protocol and the values we used in the analysis. We define α as the tag to reader $(T\Rightarrow R)$ data rate in kbps and β as the reader to tag $(R\Rightarrow T)$ data rate in kbps. Based on the parameters in Table 2 and (2), the T_p , $T_{col} - T_{no_reply} + T_3$ and T_3 in (14) can be represented as functions of $T\Rightarrow R$ data rate α and $R\Rightarrow T$ data rate β :

$$T_p = 2 \times \text{Tpri}$$
, $T_{col} - T_{no_reply} + T_3 = (39 \times M + 3) \times \text{Tpri}$ and $T_3 = 2.8 \times \text{Tari}$, where $\text{Tpri} = 1/\alpha M$ and $\text{Tari} = 1/\beta$.

Thus, we define

$$A(\alpha) = T_p = 2/\alpha M ,$$

$$B(\alpha) = T_{col} - T_{no_reply} + T_3 = (39 \times M + 3)/\alpha M ,$$

$$C(\beta) = T_3 = 2.8/\beta$$
(16)

Then, we can rewrite (14) as

$$\Delta t_{CD} \approx n_1 [3.53L \cdot A(\alpha) - 0.93 \eta_{CD} B(\alpha) - 1.60 \cdot C(\beta)]$$
 (17)

 L_{on} can be found by computing

$$\partial \Delta t_{CD} / \partial L = 0 \tag{18}$$

Using η_{CD} derived in (12), we have

$$\frac{3.53A(\alpha)}{0.93B(\alpha)} \times L^8 - 0.4714 \times L^6 - 0.2192 \times 2L^5$$

$$-0.1222 \times 3L^4 - 0.0767 \times 4L^3 - 0.0514 \times 5L^2$$

$$-0.035 \times 6L - 0.0235 \times 7 = 0$$
(19)

From (19), it can be seen that the solution for L is determined by the $T\Rightarrow R$ data rate— α and the number of subcarrier cycles per symbol in $T\Rightarrow R$ link—M [Figure 2]. With given α and M, the optimal number of time slots $L_{op}=round$ (L) can be found by computing (19). By computing (19) with a and M given in Table 2, we found that L_{op} remains the same for different α , but varies according to M. Table 3 shows the optimal number of time slots L_{op} computed for different values of M.

To verify the results in Table 3, we calculated Δt_{CD} based on (17) according to different number of time slots L to find out L_{op} . The number of tags identified is 200 (n_1 = 200), β is 40 kbps and α is 75 kbps. Δt_{CD} , according to the number of time slots L, ranging from 2 to 16, is calculated for each case of M. Figure 5 plots the calculation results. In Figure 5, for each case of M, the position of L_{op} which maximizes $|\Delta t_{CD}|$ is marked. As can be seen, the optimal number of time slots L_{op} is the same with the computation result in Table 3. The same results are obtained for different data rates α but not shown here.

4. Performance Evaluation Results

In this section, we present numerical results of the analysis in the last section and compare them with simulation results. We also compare the performance of the proposed method with the conventional methods. The simulation model is based on the Gen2 protocol and the system parameters are specified in Table 2. We set the $R \Rightarrow T$ data rate β as 40 kbps, the $T \Rightarrow R$ data rate α as 75 kbps and the number of subcarriers per symbol M as 4.

Table 2: System parameters used in the analysis

Parameters	Descriptions	Values
Tari	Reference time interval for a data-0 in $R \Rightarrow T$ link	25 μs
β	$R \Rightarrow T$ link data rate	1/Tari=40 kbps
α	$T \Rightarrow R$ link data rate	10~160 kbps
M	Number of subcarrier cycles per symbol in $T \Rightarrow R$ link	1, 2, 4, 8
LF	$T \Rightarrow R$ link frequency	$\alpha \times M$
Tpri	Pulse-repetition interval	1/LF
$T_{\rm p}$	Interval of a time slot for a pulse in the proposed scheme	2×Tpri
T_1	Time from reader transmit to tag response	2.8×Tari
T ₂	Time from tag response to reader transmit	3×Tpri
T ₃	Time a reader waits after $T_{\rm l}$, before another command	2.8 Tari
T_{Query}	Time of Query command in $R \Rightarrow T$ link	868.5 μs
T_{QueryRep}	Time of QueryRep command in $R \Rightarrow T$ link	207.5 μs
T _{RN16}	Time of RN16 reply in $T \Rightarrow R$ link	39×M×Tpri
T_{EPC}	Time of EPC reply in $T \Rightarrow R$ link	151×M×Tpri

Table 3: Optimal number of time slots L_{op} according to α and M

α	М	L	L _{op}
10~160 kbps	1	2.1225	2
	2	2.7413	3
	4	3.6346	4
	8	4.9119	5

Figure 6 shows the identification time for 200 tags using the proposed CD scheme with various values of time slots L ranging from 2 to 16. The analytical results are based on (3), (12) and (13). As can be seen from Figure 6, (1) the analytical and simulation results are close to each other; (2) the identification time reaches the minimum at L=4 such that it verifies the analytical result in the previous section. The same results are observed for different number of tags but not listed here due to limitation of space.

Figure 7 shows the performance comparison between the proposed CD method and the conventional methods in terms of identification time for different number of tags. As can be seen, an improvement of identification speed is achieved by using the proposed method. As shown in Figure 7, when identifying 300 tags, the conventional CD method in [24] costs 1.45 sec, [25] costs 1.39 sec, [26] costs 1.50 sec, while the proposed method costs 1.28 sec.

Figure 8 shows the performance evaluation result in AWGN conditions. We compare the total number of slots ($n_0 + n_1 + n_2$) used to identify the same number of tags. An average SNR ranging from 10 to 14 dB is selected to identify 200 tags. As can be seen from Figure 8, when the channel suffers an increasing level of noise, the

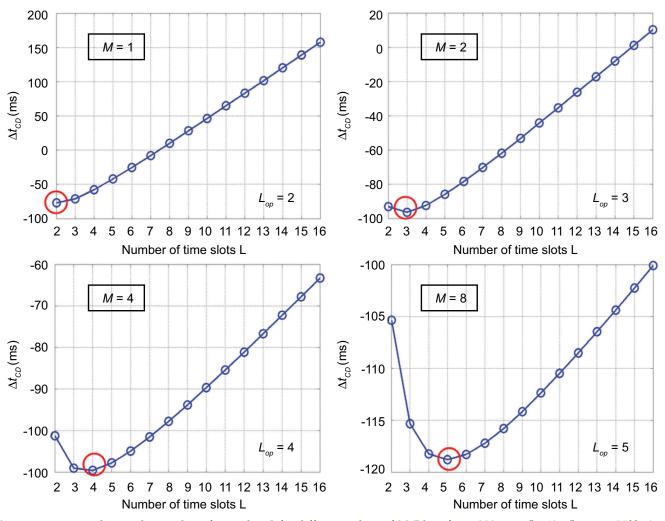


Figure 5: Δt_{CD} according to the number of time slots *L* for different values of *M* (Identifying 200 tags, β = 40 $\kappa\beta\pi\sigma$, α = 75 kbps).

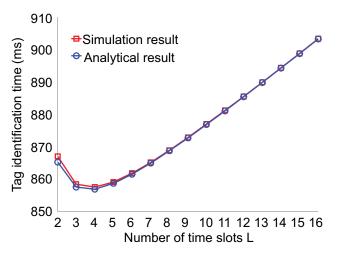


Figure 6: Tag identification time for different number of time slots *L* by using the proposed CD method to identify 200 tags.

conventional CD methods require increasing more slots than the proposed method, since the conventional methods have difficulty in distinguishing a collision from single reply corrupted by noise.

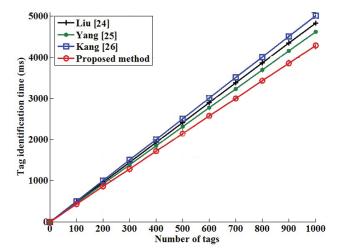


Figure 7: Tag identification time comparison (in noise free condition, $L = L_{on} = 4$ in the proposed CD method).

5. FPGA Prototype Implementation

For implementing the proposed scheme, a pulse generation function is required to implement in the tag.

Figure 9 shows the block diagram of the proposed tag chip with emphasis on baseband processor. As shown in Figure 9, the chip consists of the following three main sections: Analog front end, baseband processor and electronically erasable programmable read-only memory (EEPROM). The analog front end converts RF power received by the antenna into DC voltage which provides power supply for the rest circuits in the tag. In addition, the analog front end is responsible for receiving signals from the reader, demodulating and digitalizing the data to be used by the baseband processor. It also modulates the tag replies and transmits them to the reader [31]. The baseband processor is responsible for decoding and interpreting the incoming commands from the reader, managing the communication procedure and generating replies for the received commands. The replies generated by the baseband processor are delivered to the analog front end for sending to the reader. The proposed scheme requires no additional

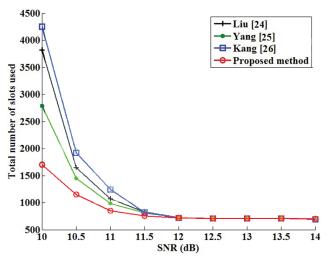


Figure 8: Total number of slots used to identify 200 tags (in AWGN conditions).

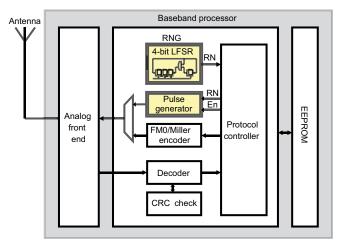


Figure 9: Block diagram of the proposed RFID tag chip with emphasis on the baseband processor.

hardware support and can be implemented by using digital logic in the baseband processor.

The baseband processor consists of several modules. The decoder module decodes the incoming data from the analog front end and extracts information included in the received data. The cyclic redundancy check (CRC) check module checks the CRC code in the decoded data for error checking. The encoder module encodes the tag reply data in either FM0 or Miller format and delivers it to the analog front end. The protocol controller is responsible for managing the whole system, deciding what operations should be done, when and what replies should be sent according to the Gen2 protocol.

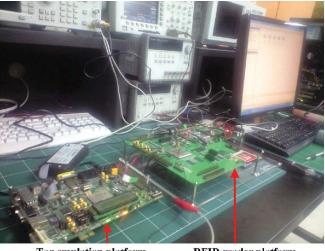
The proposed pulse generation function can be implemented by adding a random number generator (RNG) and a pulse generator to the original baseband processor, as shown in Figure 9. The RNG is used for generating a 4-bit random number (RN) that decides the time slot in which the pulse will be generated. Though the protocol controller includes a unit for generating random numbers for collision arbitration according to the Gen2 protocol, this unit is generally used for generating a 16-bit RN. Based on the analytical results in previous sections, a maximum number of 16 time slots for pulse generation are enough for achieving optimal performance for the proposed CD scheme; therefore, an independent 4-bit RNG is proposed.

The RNG can be realized by using a 4-bit linear feedback shift register (LFSR) to generate pseudorandom sequences. The LFSR is very fast and efficient for implementation. On the other hand, the LFSR follows the same implementation structure as the CRC check function, these make the LFSR an ideal choice for realizing the RNG [32]. The pulse generator can be implemented by simple digital logic to generate a pulse which is the same as one subcarrier generated by the FM0/Miller encoder. The RN generated by the RNG is sent to the pulse generator to generate a pulse in the RN-determined time slot ahead of the encoded data signal from the FM0/Miller encoder. The pulse and the encoded data signal will then be delivered to the analog front end. The protocol controller can enable or disable the pulse generation function by using a control signal "En," as shown in Figure 9. When the pulse generation function is disabled, the tag operates in conventional mode. This dual mode architecture enables the tag to be used in a wide range of conditions and facilitates performance evaluation for the tag.

Based on the proposed architecture, the baseband processor of the tag was implemented in a Xilinx Virtex-5XC5VSX50T FPGA on a Xilinx ML506 evaluation platform for functional verification. An RFID reader platform based on a Xilinx Virtex-5XC5VLX330 FPGA,

prototyped by Samsung Techwin Co., Ltd. and RFID Research Center in Hanyang University, was used to test the functionalities of the implemented tag.

Figure 10 shows the experimental setup for functional verification of the tag FPGA prototype. Aiming to test the baseband processor of the tag, the reader and tag platforms were connected by wire to enable communications using baseband signals. As shown in Figure 10, the reader sends command signals to the tag to trigger the tag operation. Then, the reader receives tag reply signals and decodes them. Measurements were done by a logic analyzer (Agilent 16903A) and the measurement results are shown in Figure 11.



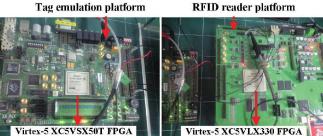


Figure 10: Experimental setup for functional verification.

In Figure 11, waveform 1 and 2 are the reader command signal and the tag reply signal, respectively. A Query command is sent by the reader to the tag at a data rate of 40 kbps. Then, an RN16 signal, which contains a pulse in the beginning part of the signal, is sent back by the tag to the reader at a data rate of 75 kbps in Miller-4 encoding format. The received RN16 signal is filtered by a digital bandpass filter in the reader for noise cancellation (waveform 3) and then decoded by the reader. Then, the reader sends an ACK command to the tag requiring its EPC. After confirming the RN16 data included in the ACK command, the tag sends an EPC signal to the reader. Waveform 4 and 5 correspond to the decoded data and data synchronization signal of the tag reply signal, respectively. Waveform 6 is a detailed view of the filtered RN16 signal shown in waveform 3. It can be seen that the proposed pulse is successfully generated ahead of the original RN16 signal. The waveform of the pulse is the same as one subcarrier in the RN16 signal. Waveform 7, 8 and 9 show the pulse detection process by using a square-law detector implemented in the reader. The samples of the pulse are saved in a register (waveform 7) and then integrated. The integration result (waveform 9) is compared with a threshold to make a decision (waveform 8). Measurement results show that the tag reply signal is correct and can be successfully decoded by the reader.

Table 4 lists the logic gate counts required for implementing the tag chip with the proposed baseband processor. The synthesis report is generated by the Synopsys Design Compiler based on the TSMC 0.18 μm technology. The implementation result is compared with that of a conventional Gen2 tag chip fabricated by Samsung Techwin Co., Ltd and RFID Research Center in Hanyang University using the same 0.18 μm technology [33]. From Table 4,

Table 4: Compiler synthesis report

	Logic gate count
Conventional tag [30]	9818
Proposed tag	9926

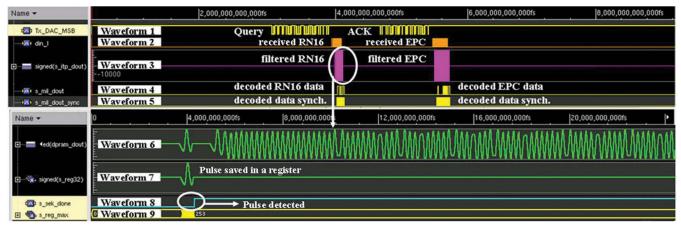


Figure 11: Measured signal waveforms from the prototyped tag FPGA platform.

it can be seen that the additional number of logic gates required for implementing the proposed chip is only 1% of the conventional chip. This slight additional cost of logic gates shows that the proposed tag baseband architecture does not increase much complexity of the tag chip and is feasible for practical implementation.

6. Inventory Procedure when the Proposed Tags are Mixed with Conventional Tags

In this part, we discuss the issue about how to separately inventory the proposed tags when they are mixed with conventional tags. Since the conventional tags currently being widely used could not support the proposed CD scheme, we propose an inventory procedure which could be used to separate the proposed tags from the conventional tags when they are mixed together. First, let us review the existing operation procedure defined in the Gen2 protocol to inventory a group of tags.

As shown in Figure 12, the reader inventories a group

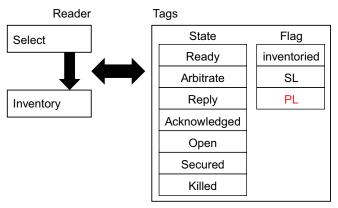


Figure 12: Operation process to inventory a group of tags.

of tags using two basic operation processes which are select and inventory. The select process allows the reader to select a tag population for inventory prior to inventory process. The reader may use one or more Select commands to select a particular tag population based on user-defined criteria. For example, the Select command can assert or deassert a tag's SL flag, or it can set a tag's inventoried flag to A or B. The SL flag and inventoried flag, as shown in Figure 12, are two existing flags in the tags which can be used for tag selection [27]. The operation of the Select command is defined by the parameters in the command. The specific parameters in the Select command are shown in Table 5. As can be seen, the "Target" and "Action" parameters indicate whether and how the Select command modifies a tag's SL or inventoried flag.

After selecting a group of tags for inventory, the selected groups of tags participate in the inventory process. The inventory process includes a set of inventory commands like *Query, QueryAdjust* and *QueryRep*, which have been described in detail in Section 2. Here, we put emphasis on the *Query* command since the *Query* command initiates the inventory process and decides which tags participate in the process. The specific parameters in the *Query* command are shown in Table 6. As can be seen from the "Sel" and "Target" parameters in the *Query* command, a *Query* command can use SL and inventoried flags to decide which tags participate in the inventory process. For example, the reader can inventory SL or ~ SL tags, or it can choose to ignore the SL flag entirely [27].

Based on the existing select and inventory process, we propose an inventory procedure to separate the proposed tags from the conventional tags. The procedure could

Table 5: Select command

	Command	Target	Action	MemBank	Pointer	Length	Mask	Truncate	CRC-16
# of bits	4	3	3	2	EBV	8	Variable	1	16
Description	1010	000: Inventoried (S0)	See Table 6	00: RFU	Starting	Mask length (bits)	Mask value	0: Disable truncation	
		001: Inventoried (S1)		01: EPC	Mask address			1: Enable truncation	
		010: Inventoried (S2)		10: TID					
		011: Inventoried (S3)		11: User					
		100: SL							
		101: RFU→ PL							
		110: RFU							
		111: RFU							

Table 6: Query command

	Command	DR	М	TRext	Sel	Session	Target	Q	CRC-5
# of bits	4	1	2	1	2	2	1	4	5
Description	1000	0: DR=8	00: M=1	0: No pilot tone	00: All	00: S0	0: A	0-15	
		1: DR=64/3	01: M=2	1: Use pilot tone	01: 	01: \$1	1: B		
			10: M=4		10: ∼SL	10: \$2			
			11: M=8		11: SL	11: \$3			

be achieved by adding a PL flag in additional to the existing SL and inventoried flags in the tag (as shown in Figure 12) and a slight updating of the existing Select and Query commands. The purpose of the supposed PL flag is that the reader could uniquely select the proposed tags for inventory by setting the "Target" parameter in the Select command and the "Sel" parameter in the Query command to PL, as shown in Tables 5 and 6. And, a corresponding tag response action to the updated Select command is supposed to be "assert PL" instead of the original "do nothing," as shown in Table 7. Since the proposed updating fields in the original reader commands (as shown in Tables 5 and 6) and tag response [Table 7] are not under use currently, the proposed update is feasible and will not affect the operation procedure of the conventional systems.

The proposed inventory procedure to separate the proposed tags from the conventional tags is illustrated in Figure 13. As can be seen from Figure 13, the select and inventory processes are identical to the conventional processes, except the usage of the PL flag proposed uniquely for the proposed tags.

7. Conclusion

In this paper, we proposed a scheme to improve the CD

Table 7: Tag response to *Select* command according to "Action" parameter

Action	Matching	Non-Matching		
000	assert SL or inventoried ->A	deassert SL or inventoried -> B		
001	assert SL or inventoried ->A	do nothing		
010	do nothing → assert PL	deassert SL or inventoried -> B		
•••				
111	do nothing	negate SL or (A->B, B->A)		

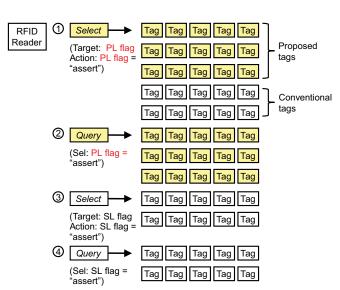


Figure 13: Inventory procedure to separate proposed tags from conventional tags when they are mixed together.

efficiency of the Gen2 RFID systems. An improvement in tag identification speed compared with the conventional methods was verified by simulation. We also discussed development issues for a Gen2 RFID tag to support the implementation of the proposed scheme. Functional measurement and synthesis result of the implemented baseband processor validates the practical feasibility to fabricate the tag chip. Moreover, using the proposed scheme, we are able to detect the number of collided tags in a collision that will help in estimating the unidentified tag population. Combining the proposed scheme with tag estimation method to maximize system efficiency is our future research subject.

8. Acknowledgments

This work was supported by Korea Evaluation Institute of Industrial Technology (KEIT), under the R and D support program of Ministry of Knowledge Economy, Korea.

References

- Z.H. Tee, L.M. Ang, and K.P. Seng, "Smart guide system to assist visually impaired people in an indoor environment", IETE Technical Review, Vol. 27, no. 6, pp. 455-64, Nov-Dec 2010.
- D. Guan, T. Ma, W. Yuan, Y.K. Lee, and A.M. Jehad Sarkar, "Review of sensor-based activity recognition systems", IETE Technical Review, Vol. 28, no. 5, pp. 418-33, Sep-Oct 2011.
- M.S.I. Sayeed, Y.S. Kim, H. Yang, and J.C. Yook, "A solution to the RFID reader interference problem using adaptive beam-forming approach", IETE Technical Review, Vol. 28, no. 1, pp. 17-28, Jan-Feb 2011.
- J.I. Capetanakis, "Tree algorithms for packet broadcast channels", IEEE Trans. Inform. Theory, Vol. 25, pp. 505-515, Sept. 1979.
- D.R. Hush, and C. Wood, "Analysis of tree algorithms for RFID arbitration", IEEE International Symposium on Information Theory, Cambridge, MA, USA, pp. 107, 1998.
- J. Myung, and W. Lee, "Adaptive binary splitting: A RFID tag collision arbitration protocol for tag identification", Mobile Networks and Applications, Vol. 11, no. 5, pp. 711-22, 2006.
- J.H. Choi, D. Lee, and H. Lee, "Bi-slotted tree based anti-collision protocols for fast tag identification in RFID systems", IEEE Commun. Lett., Vol. 10, no. 12, pp. 861-3, 2006.
- 8. J. Ryu, H. Lee, Y. Seok, T. Kwon, and Y. Choi, "A hybrid query tree protocol for tag collision arbitration in RFID systems", in Proc. IEEE International Conference on Communications (ICC'07), Seoul, Korea, pp. 5981-6, 2007.
- 9. Y. Cui, and Y. Zhao, "Performance evaluation of a multi-branch tree algorithm in RFID", IEEE Trans. Commun., Vol. 58, no. 5, pp. 1356-64, 2010.
- C.N. Yang, Y.C. Kun, C.Y. Chiu, and Y.Y. Chu, "A new adaptive query tree on resolving RFID tag collision", 2010 IEEE International Conference on RFID-Technology and Applications (RFID-TA), Guangzhou, China, pp. 153-8, 2010.
- H. Vogt, "Efficient object identification with passive RFID tags", First International Conference on Pervasive Computing, pp. 98-113, 2002.
- S.R. Lee, S.D. Joo, and C.W. Lee, "An enhanced dynamic framed slotted ALOHA algorithm for RFID tag identification", in Proc. The Second Annual International Conference on Mobile and Ubiquitous Systems: Networking and Services (MobiQuitous'05), San Diego, CA, pp. 166-72, 2005.
- C. Floerkemeier, "Bayesian transmission strategy for framed ALOHA based RFID protocols", IEEE International Conference

- on RFID, Grapevine, TX, pp. 228-35, 2007.
- C. Lee, and H. Cho, "An adaptive RFID anti-collision algorithm based on dynamic framed ALOHA", IEICE Trans. Commun., Vol. E91-B, no. 2, pp. 641-5, 2008.
- Y.J. Lee, D.K. Kwon, and H.N. Kim, "Collision arbitration based on different slot times for slotted-Aloha RFID systems", IEICE Trans. Commun., Vol. E91-B, no. 5, pp. 1416-22, 2008.
- L. Zhu, and T.S.P. Yum, "Optimal framed Aloha based anti-collision algorithms for RFID systems", IEEE Trans. Commun., Vol. 58, no. 12, pp. 3583-92, 2010.
- H. Wu, and Y. Zeng, "Efficient framed slotted Aloha protocol for RFID tag anticollision", IEEE Trans. Autom. Sci. Eng., Vol. 8, no. 3, pp. 581-8, 2011.
- F.C. Schoute, "Dynamic frame length ALOHA", IEEE Trans. Commun., Vol. 31, no. 4, pp. 565-8, 1983.
- S.S. Choi, and S. Kim, "A dynamic framed slotted ALOHA algorithm using collision factor for RFID identification", IEICE Trans. Commun., Vol. E92, no. 3, pp. 1023-6, 2009.
- J.R. Cha, and J.H. Kim, "Dynamic framed slotted ALOHA algorithms using fast tag estimation method for RFID system", IEEE Consumer Communications and Networking Conference, Las Vegas, USA, pp. 768-72, 2006.
- W.T. Chen, "An accurate tag estimate method for improving the performance of an RFID anticollision algorithm based on dynamic frame length ALOHA", IEEE Trans. Autom. Sci. Eng., Vol. 6, no. 1, pp. 9-15, 2009.
- J.B. Eom, and T.J. Lee, "Accurate tag estimation for dynamic framed-slotted ALOHA in RFID systems", IEEE Commun. Lett., Vol. 14, no. 1, pp. 60-2, 2010.
- R.S. Khasgiwale, R.U. Adyanthaya, and D.W. Engels, "Extracting information from tag collisions", IEEE International Conference on RFID, Orlando, FL, pp. 131-8, 2009.

- D. Liu, Z. Wang, J. Tan, H. Min, and J. Wang, "ALOHA algorithm considering the slot duration difference in RFID system", IEEE International Conference on RFID, Orlando, FL, pp. 56-63, 2009.
- L. Yang, J. Han, Y. Qi, C. Wang, Y. Liu, and Y. Cheng, et al., "Revisting tag collision problem in RFID systems", 2010 39th International Conference on Parallel Processing, San Diego, USA, pp. 178-87, 2010.
- L. Kang, K. Wu, J. Zhang, H. Tan, and L. Ni, "DDC: A novel scheme to directly decode the collisions in UHF RFID systems", IEEE Trans. Parallel Distrib. Syst., accepted for future publication, Vol. 23, pp. 1-8, 2012.
- EPCglobal, "EPC radio-frequency identity protocols class-1 generation-2 UHF RFID protocol for communications at 860-960 MHz version 1.2.0," 2008.
- W.F. Lo, and H.T. Mouftah, "Collision detection and multitone tree search for multiple-access protocols on radio channels", IEEE J. Sel. Areas Commun., Vol. 5, no. 6, pp. 1035-40, 1987.
- Y. Maguire, and R. Pappu, "An optimal Q-algorithm for the ISO 18000-6C RFID protocol", IEEE Trans. Autom. Sci. Eng., Vol. 6, no. 1, pp. 16-24, 2009.
- 30. B. Li, Y. Yang, and J. Wang, "Anti-collision issue analysis in Gen2 protocol", Auto-ID Labs White Paper, WP-HARDWARE-047,
- J. Melia-Segui, J. Garcia-Alfaro, and J. Herrera-Joancomarti, "Analysis and improvement of a pseudorandom number generator for EPC Gen2 tags", 14th International Conference on Financial Cryptography and Data Security, Spain, pp. 34-46, Jan. 2010.
- K.Y. Jeon, and S.H. Cho, "A RFID EPC C1 Gen2 system with channel coding capability in AWGN noise environments", IEICE Trans. Commun., Vol. E92-B, no. 2, pp. 608-11, Feb. 2009.

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